

Maximum Correntropy Filtering for Complex Networks With Uncertain Dynamical Bias: Enabling Component-Wise Event-Triggered Transmission

Weihao Song, Zidong Wang, Zhongkui Li, Qing-Long Han and Dong Yue

Abstract—This paper is concerned with the maximum correntropy filtering problem for a class of nonlinear complex networks subject to non-Gaussian noises and uncertain dynamical bias. With aim to utilize the constrained network bandwidth and energy resources in an efficient way, a component-wise dynamic event-triggered transmission (DETT) protocol is adopted to ensure that each sensor component independently determines the time instant for transmitting data according to the individual triggering condition. The principal purpose of the addressed problem is to put forward a dynamic event-triggered recursive filtering scheme under the maximum correntropy criterion such that the effects of the non-Gaussian noises can be attenuated. In doing so, a novel correntropy-based performance index (CBPI) is first proposed to reflect the impacts from the component-wise DETT mechanism, the system nonlinearity and the uncertain dynamical bias. The CBPI is parameterized by deriving upper bounds on the one-step prediction error covariance and the equivalent noise covariance. Subsequently, the filter gain matrix is designed by means of maximizing the proposed CBPI. Finally, an illustrative example is provided to substantiate the feasibility and effectiveness of the developed maximum correntropy filtering scheme.

Index Terms—Complex networks, maximum correntropy filtering, non-Gaussian noises, dynamical bias, dynamic event-triggered protocol.

I. INTRODUCTION

Along with the burgeoning information technologies with applications in network sciences, the past several decades have witnessed a steadily growing research interest in complex networks that are capable of modeling a wide variety of

real-world systems. Some representative complex networks include, but are not limited to, social networks, airport networks, World Wide Web, electric power grids, metabolic networks, neural networks, and genetic regulatory networks [6], [33], [36], [48]. Roughly speaking, a typical complex network is constituted by a group of interplayed dynamical nodes where each individual node shares its local information with its neighbors with aim to accomplish certain missions in a collaborative manner.

Inspired by their theoretical significance and extensive applicability, the dynamical analysis issues for complex networks have drawn a large amount of research attention leading to many excellent results reported in the literature, see e.g. [35], [42]. For example, the output synchronization issue has been considered in [41] for uncertain general complex networks based on the neural sliding-mode pinning control strategy. In [29], the finite/fixed-time synchronization issue has been tackled for complex networks subject to stochastic disturbances by means of a unified pinning controller with regulatable power parameters.

There has been an ever-increasing demand for acquiring the states of underlying system plants in general practice such as monitoring and control of process engineering. Correspondingly, the state estimation (or filtering) issues have become a research focus for several decades and a great number of effective estimation/filtering schemes have been put forward for various systems with different performance requirements, see e.g. [3], [4], [23], [49] and the references therein. In particular, in the context of complex networks, the relevant results have been obtained based primarily on the Kalman-like (or extended-Kalman-like) filtering approach [11], [25], the H_∞ filtering approach [43], [50], and the set-membership filtering approach [27], [39]. It should be noted that most existing results are predominantly applicable to Gaussian noises as well as energy- or norm-bounded noises.

In engineering practice, owing for example to the signal reflection and impulsive electromagnetic interference, non-Gaussian noises are frequently encountered in many practical scenarios such as underwater acoustic localization and ballistic target tracking [18], [21]. As such, much effort has been devoted to the treatment of non-Gaussian noises and several effective strategies have been developed, for instance, the celebrated particle filter [2], the Gaussian sum filter [1] and the variational Bayesian filter [57], to name just a few. Among others, the maximum correntropy filter [5], [16] has stood out as an easy-to-implement yet efficacious approach to handle

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the impulsive non-Gaussian noises and measurement outliers, where the core idea lies in the fact that the correntropy, as an information theoretic criterion, is able to reflect the higher-order statistics of the probability density function [28].

Due to its distinguishing merits of flexibility (in filter design), robustness (against outliers) and computational cost, the maximum correntropy filtering (MCF) scheme has begun to stir some initial attention and several excellent results have been reported in recent years [38], [40], [45], [46]. For example, a maximum correntropy Kalman filter, combined with the fixed-point method, has been developed in [5] for linear systems. In [22], the Chandrasekhar-type recursion has been derived for the maximum correntropy Kalman filtering scheme proposed in [16]. It is worth mentioning that limited work has been done towards the complex networks so far. One of the few available results is the maximum correntropy filter developed for *linear* complex networks undergoing random occurring cyber-attacks [37]. When it comes to *nonlinear* complex networks with engineering-oriented complexities, the corresponding MCF issue has not been thoroughly studied yet.

It has now been widely recognized that, in most practical applications, the system dynamics and/or sensor measurements might be affected by not only the modeled noises but also the unknown inputs (e.g. unknown disturbances, unmodeled dynamics, component faults and imperfect calibrations) [17], [19], [34]. Such unknown inputs, if not delicately tackled, are likely to incur severe performance deterioration and even failure of certain tasks (e.g. target tracking). In this regard, noticeable attention has been paid to the investigation on joint state and input estimation issues, see [9], [52], [53] and the references therein. It should be noted that, a kind of unknown inputs, namely, stochastic bias, has attracted a particular research interest in the literature [10], [15], [19], where a prerequisite for most existing results is that the bias dynamics can be precisely determined, which might not be the case in practice for many reasons such as environment-induced parameter perturbations. Taking the typical target tracking problem as an example, the actuators (e.g. rudders) of the targets of interest might undergo the uncertain dynamical bias in a rather complicated environment as a result of abrupt external disturbances, random operation errors and electromagnetic interferences [44]. Consequently, it makes practical sense to investigate the filtering problem for complex networks subject to *uncertain* dynamical bias.

With the nowadays popularity of networked systems, the network-induced imperfections (e.g. network traffic congestion) has grabbed considerable research attention from both academia and industry, see e.g. [14], [27], [55], and [54] for a recent survey. In particular, the so-called event-triggered transmission protocol has proven to be an effective approach to mitigating unnecessary network traffic, thereby extenuating the occurrence of network-induced imperfections [8], [20], [26], [30], [47], [56]. In comparison with its static counterpart, the dynamic event-triggered transmission (DETT) protocol has better potentials in economizing the limited network resources since its threshold can be dynamically adjusted, see e.g. [7], [12], [13], [24], [32] for some DETT-based results. It should be stressed that in most existing results, the measured outputs

have been assumed to share a common triggering condition, which is sometimes unreasonable since the output magnitude and the required update frequency of each individual component might be different. Towards this direction, a component-wise DETT protocol has been proposed in investigating the state estimation problem for linear systems with non-uniform sampling, which allows for different triggering conditions for different components [58]. Nevertheless, the corresponding results for nonlinear complex networks have been very few (if not none), let alone the consideration of uncertain dynamical bias and maximum correntropy criterion.

Motivated by the aforementioned discussions, in this paper, we endeavor to handle the MCF problem for a class of nonlinear complex networks in the presence of non-Gaussian noises and uncertain dynamical bias under component-wise DETT protocol. In doing so, we foresee the following three essential challenges: 1) how to establish an appropriate model for complex networks with uncertain dynamical bias and non-Gaussian noises, where the measurement transmissions are scheduled by the component-wise DETT protocol? 2) how to design a correntropy-based performance index (CBPI) that quantifies the joint influence from the considered system complexities and transmission scheduling mechanism? and 3) how to deal with the analytical complexity induced by the component-wise DETT protocol? As such, the primary objective of this current investigation is to overcome the challenges listed above.

The major contributions of this paper are highlighted from the following three aspects: 1) the MCF problem is, for the first time, investigated for nonlinear and non-Gaussian complex networks subject to uncertain dynamical bias under the component-wise DETT protocol; 2) a novel CBPI is established, by resorting to the upper bounds on the prediction error covariance and the equivalent noise covariance, to take into account the impacts from uncertain dynamical bias and component-wise DETT protocol; and 3) an easy-to-implement algorithm is developed to realize the resource-saving filter, which is suitable for online computations with desired robustness against non-Gaussian noises.

The rest of this paper is outlined as follows. Section II specifies the problem under investigation, and presents the component-wise DETT mechanism as well as the MCF scheme. In Section III, the proposed CBPI is explicitly expressed by determining two weighted matrices, based on which the filter gain matrix is designed. In Section IV, simulation results are exhibited to verify the effectiveness of the proposed filtering algorithm. Finally, some concluding remarks are summarized in Section V.

Notation. The notations adopted throughout this paper are fairly standard. \mathbb{R}^n denotes the n -dimensional Euclidean vector space. The superscripts T and -1 stands for, respectively, the operation of transpose and inverse. $\|x\|$ and $\|x\|_A \triangleq (x^T A x)^{1/2}$ represent, respectively, the Euclidean norm and the weighted norm of a vector $x \in \mathbb{R}^n$, where A is a positive definite matrix. $\mathbb{E}\{\cdot\}$ refers to the mathematical expectation operator. $\text{diag}\{\cdot\cdot\cdot\}$ denotes a block-diagonal matrix. 0 and I denote, respectively, the zero matrix and the identity matrix of compatible dimensions. Other notations will be interpreted as

the need arises.

II. PROBLEM FORMULATION

Consider the following class of complex networks composed of N coupled nodes with dynamical biases:

$$\bar{x}_{i,s+1} = \bar{A}_{i,s}\bar{x}_{i,s} + \bar{f}_s(\bar{x}_{i,s}) + \sum_{j=1}^N d_{ij}\bar{\Gamma}\bar{x}_{j,s} + B_{i,s}z_{i,s} + \zeta_{i,s} \quad (1)$$

where $\bar{x}_{i,s} \in \mathbb{R}^{\bar{n}}$ denotes the state vector of the i -th node at time instant s , $D = [d_{ij}]_{N \times N}$ represents the coupled configuration matrix and $\bar{\Gamma} = \text{diag}\{\gamma_1, \gamma_2, \dots, \gamma_{\bar{n}}\}$ stands for the inner coupling matrix. $\bar{A}_{i,s}$ and $B_{i,s}$ are the known matrices with appropriate dimensions. $\zeta_{i,s}$ refers to the process noise with zero mean and covariance $Q_{i,s} > 0$. $z_{i,s} \in \mathbb{R}^k$ is the random bias with the following dynamics

$$z_{i,s+1} = (G_{i,s} + \Delta G_{i,s})z_{i,s} + \eta_{i,s} \quad (2)$$

where $G_{i,s}$ represents the known bias transition matrix and $\Delta G_{i,s}$ implies the perturbation term satisfying

$$\begin{aligned} \mathbb{E}\{\Delta G_{i,s+1}\} &= 0, \\ \mathbb{E}\{(\Delta G_{i,s+1})(\Delta G_{i,s+1}^T)\} &\leq \tau_i I \end{aligned}$$

where τ_i is a given positive scalar, $\eta_{i,s}$ denotes the noise with zero mean and covariance $S_{i,s} > 0$. The known nonlinear function $\bar{f}_s(\cdot)$ satisfies $\bar{f}_s(0) = 0$ and

$$\|\bar{f}_s(u) - \bar{f}_s(v) - \bar{F}_s(u-v)\| \leq \kappa_s \|u-v\| \quad (3)$$

where $u, v \in \mathbb{R}^{\bar{n}}$, \bar{F}_s is a known matrix, and $\kappa_s \geq 0$ is a known scalar.

The measurement model of the i -th node is characterized by

$$y_{i,s} = \bar{C}_{i,s}\bar{x}_{i,s} + \nu_{i,s} \quad (4)$$

where $y_{i,s} \in \mathbb{R}^m$ denotes the measurement output and $\nu_{i,s}$ stands for the measurement noise with zero mean and covariance $R_{i,s} > 0$. Throughout this paper, we assume that $\zeta_{i,s}$, $\eta_{i,s}$, $\nu_{i,s}$, $\bar{x}_{i,0}$, $z_{i,0}$ and the elements of $\Delta G_{i,s}$ are mutually independent in i and s .

Denote

$$\begin{aligned} x_{i,s} &= [\bar{x}_{i,s}^T \quad z_{i,s}^T]^T, A_{i,s} = \begin{bmatrix} \bar{A}_{i,s} & B_{i,s} \\ 0 & G_{i,s} \end{bmatrix}, \\ \Delta A_{i,s} &= \begin{bmatrix} 0 & 0 \\ 0 & \Delta G_{i,s} \end{bmatrix}, \Gamma = \begin{bmatrix} \bar{\Gamma} & 0 \\ 0 & 0 \end{bmatrix}, \\ f_s(x_{i,s}) &= [\bar{f}_s(\bar{x}_{i,s})^T \quad 0]^T, \omega_{i,s} = [\zeta_{i,s}^T \quad \eta_{i,s}^T]^T, \\ C_{i,s} &= [\bar{C}_{i,s} \quad 0]. \end{aligned}$$

Then, the augmented dynamics can be obtained as follows:

$$\begin{cases} x_{i,s+1} = (A_{i,s} + \Delta A_{i,s})x_{i,s} + f_s(x_{i,s}) \\ \quad + \sum_{j=1}^N d_{ij}\Gamma x_{j,s} + \omega_{i,s} \\ y_{i,s} = C_{i,s}x_{i,s} + \nu_{i,s}. \end{cases} \quad (5)$$

A. Component-wise dynamic event-triggered mechanism

In this paper, to ameliorate the utilization efficiency of the limited network bandwidth and energy resources, the component-wise DETT mechanism is employed to schedule the process of data transmission. To this end, the measurements of the i -th node are rewritten by

$$y_{i,s} = [y_{i,1,s} \quad y_{i,2,s} \quad \dots \quad y_{i,m,s}]^T$$

where $y_{i,l,s}$ ($l = 1, 2, \dots, m$) denotes the measurement output of the l -th sensor component for the i -th node.

Define the event generator function $g(r_{i,l,s}, \pi_{i,l}, \rho_{i,l}, \xi_{i,l,s})$ as follows [7], [32]:

$$g(r_{i,l,s}, \pi_{i,l}, \rho_{i,l}, \xi_{i,l,s}) = r_{i,l,s}^2 - \pi_{i,l} - \frac{1}{\rho_{i,l}} \xi_{i,l,s} \quad (6)$$

where $r_{i,l,s} = y_{i,l,s} - \check{y}_{i,l,s}$, and $\check{y}_{i,l,s}$ refers to the transmitted measurement at latest triggering time instant. $\pi_{i,l}$ and $\rho_{i,l}$ are, respectively, the prescribed triggering threshold and adjustable parameter. $\xi_{i,l,s}$ represents an auxiliary variable governed by the following dynamics

$$\xi_{i,l,s} = \delta_{i,l} \xi_{i,l,s-1} - r_{i,l,s-1}^2 + \pi_{i,l} \quad (7)$$

where $\delta_{i,l}$ denotes a predefined parameter and the initial auxiliary variable satisfies $\xi_{i,l,0} \geq 0$.

For the convenience of subsequent analysis, define an indicator variable $\lambda_{i,l,s}$ as follows:

$$\lambda_{i,l,s} = \begin{cases} 1, & g(r_{i,l,s}, \pi_{i,l}, \rho_{i,l}, \xi_{i,l,s}) > 0; \\ 0, & g(r_{i,l,s}, \pi_{i,l}, \rho_{i,l}, \xi_{i,l,s}) \leq 0. \end{cases} \quad (8)$$

Then, the available measurement at the filter end, denoted by $\bar{y}_{i,l,s}$, can be written as

$$\bar{y}_{i,l,s} = y_{i,l,s} - (1 - \lambda_{i,l,s})r_{i,l,s}. \quad (9)$$

In what follows, let us denote

$$\begin{aligned} \bar{y}_{i,s} &= [\bar{y}_{i,1,s} \quad \bar{y}_{i,2,s} \quad \dots \quad \bar{y}_{i,m,s}]^T, \\ r_{i,s} &= [r_{i,1,s} \quad r_{i,2,s} \quad \dots \quad r_{i,m,s}]^T, \\ \Lambda_{i,s} &= \text{diag}\{\lambda_{i,1,s}, \lambda_{i,2,s}, \dots, \lambda_{i,m,s}\}. \end{aligned}$$

Accordingly, the compact form of the available measurements can be reformulated by

$$\bar{y}_{i,s} = y_{i,s} - (I - \Lambda_{i,s})r_{i,s}. \quad (10)$$

B. Maximum-correntropy-based filtering scheme

In this paper, a similarity measure named correntropy is considered to enhance the robustness against the non-Gaussian noises. Compared with the widely used minimum mean square error criterion that only utilizes the second-order statistics, the utilization of correntropy can capture not only the second-order statistics but also the higher-order ones, thereby exhibiting great potentials in improving the filtering performance. More specifically, given any two scalar random variables X and Y , the correntropy $V(X, Y)$ is defined as follows [5], [28]:

$$V(X, Y) = \mathbb{E}\{\epsilon(X, Y)\} = \iint_{x,y} \epsilon(x, y) f_{X,Y}(x, y) dx dy \quad (11)$$

where $\epsilon(X, Y)$ denotes the kernel function, and $f_{X,Y}(x, y)$ is the joint probability density function of X and Y . In this paper, the kernel function is chosen to be the most popular Gaussian type, i.e.,

$$\epsilon(x, y) = G_\chi(e) = \exp\left(-\frac{e^2}{2\chi^2}\right) \quad (12)$$

where $e = x - y$ and $\chi > 0$ refers to the bandwidth of kernel size.

Considering the fact that it is usually difficult (if not impossible) to obtain the analytical expression of the joint density $f_{X,Y}(x, y)$ and only a few number of samples $\{x_i, y_i\}_{i=1}^M$ are accessible, the correntropy is approximately evaluated as follows:

$$V(X, Y) \approx \frac{1}{M} \sum_{i=1}^M \exp\left(-\frac{(x_i - y_i)^2}{2\chi^2}\right). \quad (13)$$

It is clear that in the case of Gaussian kernel, the correntropy reaches its maximum value only when $X = Y$.

In this paper, the filter for the i -th node is constructed with the following two-stage recursive form:

$$\hat{x}_{i,s|s-1} = A_{i,s-1}\hat{x}_{i,s-1} + f_{s-1}(\hat{x}_{i,s-1}) + \sum_{j=1}^N d_{ij}\Gamma\hat{x}_{j,s-1} \quad (14)$$

$$\hat{x}_{i,s} = \hat{x}_{i,s|s-1} + K_{i,s}(\bar{y}_{i,s} - C_{i,s}\hat{x}_{i,s|s-1}) \quad (15)$$

where $\hat{x}_{i,s|s-1}$ and $\hat{x}_{i,s}$ stand for, respectively, the one-step prediction and state estimate of the i -th node at time instant s , and $K_{i,s}$ denotes the filter gain matrix to be designed. Note that if the available measurements $\bar{y}_{i,s}$ are replaced by $y_{i,s}$ herein, the resultant filter is customized for the case without component-wise DETT mechanism.

In what follows, the CBPI is constructed for the purpose of designing the filter gain, i.e.,

$$J(x_{i,s}) = G_\chi\left(\|x_{i,s} - \hat{x}_{i,s|s-1}\|_{\mathcal{P}_{i,s|s-1}^{-1}}\right) + G_\chi\left(\|\bar{y}_{i,s} - C_{i,s}x_{i,s}\|_{\mathcal{R}_{i,s}^{-1}}\right) \quad (16)$$

where $\mathcal{P}_{i,s|s-1}$ and $\mathcal{R}_{i,s}$ denote the weighted matrices, which will be determined in the subsequent section.

On the basis of the maximum correntropy criterion, the desired filter gain $K_{i,s}$ can be obtained by solving the following optimization problem:

$$\hat{x}_{i,s} = \arg \max_{x_{i,s}} J(x_{i,s}). \quad (17)$$

Remark 1: It should be pointed out that, in most existing literature (e.g. [16], [46]), the one-step prediction error covariance and the measurement noise covariance are introduced as the weighted matrices in the CBPI with aim to realize the minimum-variance estimation. Nevertheless, due mainly to the existence of the component-wise DETT mechanism as well as the uncertain dynamical bias, it would be a rather challenging task to parameterize the accurate correntropy dynamics in a similar way. To this end, a novel performance index of the form (16) is proposed in this paper, where the latest transmitted measurements $\bar{y}_{i,s}$ (instead of the newly measured outputs

$y_{i,s}$) are utilized. In comparison with the standard CBPI, our proposed one with weighted matrices $\mathcal{P}_{i,s|s-1}$ and $\mathcal{R}_{i,s}$ (which are actually the corresponding upper bounds of the one-step prediction covariance and equivalent noise covariance) is able to attenuate the joint effects of the component-wise DETT mechanism and uncertain dynamical bias.

Remark 2: It is worthwhile to mention that in this paper, our attention is mainly focused on the *remote* state estimation problem, where the sensors and the remote filters are not deployed in the same positions. Consider a scenario where the small unmanned aerial vehicles are utilized as sensors to detect or track the targets of interest, and the corresponding measurement data is transmitted to the ground control station (including the filters) for further processing. To prolong the lifespan of sensors, the component-wise DETT mechanism can be exploited to select the necessary data transmissions. Meanwhile, the real-time data transmissions among remote filters can be retained to guarantee the reliable state estimation. When the energy supply of filters becomes a major concern, it is of practical significance to ameliorate the utilization efficiency of limited resources by averting frequent data transmissions among filters, which deserves further investigations.

The primary purpose of this paper is to develop a recursive filter of the structure (14)-(15) for the uncertain non-Gaussian complex networks described by (1)-(4) under the component-wise DETT mechanism (6)-(7) as well as the maximum correntropy criterion (16)-(17).

III. MAXIMUM-CORRENTROPY-BASED FILTER DESIGN AND DISCUSSION

In this section, a recursive filtering scheme is developed based on the maximum correntropy criterion. To be more specific, we first establish an explicit expression for the proposed performance index by determining the weighted matrices $\mathcal{P}_{i,s|s-1}$ and $\mathcal{R}_{i,s}$, and then design the filter gain matrix by maximizing the proposed performance index.

To begin with, let us denote

$$\begin{aligned} \tilde{x}_{i,s|s-1} &= x_{i,s} - \hat{x}_{i,s|s-1}, \quad \tilde{x}_{i,s} = x_{i,s} - \hat{x}_{i,s} \\ P_{i,s|s-1} &= \mathbb{E}\{\tilde{x}_{i,s|s-1}\tilde{x}_{i,s|s-1}^T\}, \quad P_{i,s} = \mathbb{E}\{\tilde{x}_{i,s}\tilde{x}_{i,s}^T\} \end{aligned} \quad (18)$$

where $\tilde{x}_{i,s|s-1}$ and $\tilde{x}_{i,s}$ represent, respectively, the one-step prediction error and the filtering error for the i -th node at time instant s . $P_{i,s|s-1}$ and $P_{i,s}$ denote, respectively, the corresponding one-step prediction error covariance and filtering error covariance.

According to (5), (14) and (18), the dynamics of the one-step prediction error can be given by

$$\begin{aligned} \tilde{x}_{i,s|s-1} &= A_{i,s-1}\tilde{x}_{i,s-1} + \Delta A_{i,s-1}x_{i,s-1} + \tilde{f}_{s-1}(x_{i,s-1}) \\ &\quad + \sum_{j=1}^N d_{ij}\Gamma\tilde{x}_{j,s-1} + \omega_{i,s-1} \end{aligned} \quad (19)$$

where $\tilde{f}_{s-1}(x_{i,s-1}) = f_{s-1}(x_{i,s-1}) - f_{s-1}(\hat{x}_{i,s-1})$.

Recalling the definition of prediction error covariance in (18) and the fact that $\omega_{i,s-1}$ is independent of $\tilde{x}_{i,s-1}$, $\tilde{x}_{j,s-1}$

and $x_{i,s-1}$, it is not difficult to obtain that

$$\begin{aligned}
& P_{i,s|s-1} \\
&= \mathbb{E}\{\tilde{x}_{i,s|s-1}\tilde{x}_{i,s|s-1}^T\} \\
&= A_{i,s-1}P_{i,s-1}A_{i,s-1}^T + \mathbb{E}\{\Delta A_{i,s-1}x_{i,s-1}x_{i,s-1}^T\Delta A_{i,s-1}^T\} \\
&\quad + \mathbb{E}\{\tilde{f}_{s-1}(x_{i,s-1})\tilde{f}_{s-1}(x_{i,s-1})^T\} + Q_{i,s-1} \\
&\quad + \mathbb{E}\left\{\left(\sum_{j=1}^N d_{ij}\Gamma\tilde{x}_{j,s-1}\right)\left(\sum_{j=1}^N d_{ij}\Gamma\tilde{x}_{j,s-1}\right)^T\right\} \\
&\quad + \mathbb{E}\{A_{i,s-1}\tilde{x}_{i,s-1}\tilde{f}_{s-1}(x_{i,s-1})^T\} + \mathbb{E}\{\tilde{f}_{s-1}(x_{i,s-1}) \\
&\quad \times \tilde{x}_{i,s-1}^T A_{i,s-1}^T\} + \mathbb{E}\left\{A_{i,s-1}\tilde{x}_{i,s-1}\left(\sum_{j=1}^N d_{ij}\Gamma\tilde{x}_{j,s-1}\right)^T\right\} \\
&\quad + \mathbb{E}\left\{\left(\sum_{j=1}^N d_{ij}\Gamma\tilde{x}_{j,s-1}\right)\tilde{x}_{i,s-1}^T A_{i,s-1}^T\right\} + \mathbb{E}\{\tilde{f}_{s-1}(x_{i,s-1}) \\
&\quad \times \left(\sum_{j=1}^N d_{ij}\Gamma\tilde{x}_{j,s-1}\right)^T\} + \mathbb{E}\left\{\left(\sum_{j=1}^N d_{ij}\Gamma\tilde{x}_{j,s-1}\right)\tilde{f}_{s-1}(x_{i,s-1})^T\right\}
\end{aligned} \tag{20}$$

where $Q_{i,s-1} = \text{diag}\{\bar{Q}_{i,s-1}, S_{i,s-1}\}$.

On the other hand, subtracting (15) from $x_{i,s}$ and using (10) lead to the following dynamics of filtering error:

$$\begin{aligned}
& \tilde{x}_{i,s} \\
&= \tilde{x}_{i,s|s-1} - K_{i,s}(y_{i,s} - (I - \Lambda_{i,s})r_{i,s} - C_{i,s}\hat{x}_{i,s|s-1}) \\
&= (I - K_{i,s}C_{i,s})\tilde{x}_{i,s|s-1} - K_{i,s}\nu_{i,s} + K_{i,s}(I - \Lambda_{i,s})r_{i,s}.
\end{aligned} \tag{21}$$

In view of (18) and (21), the filtering error covariance can be derived as follows:

$$\begin{aligned}
& P_{i,s} \\
&= \mathbb{E}\{\tilde{x}_{i,s}\tilde{x}_{i,s}^T\} \\
&= (I - K_{i,s}C_{i,s})P_{i,s|s-1}(I - K_{i,s}C_{i,s})^T + K_{i,s}R_{i,s}K_{i,s}^T \\
&\quad + \mathbb{E}\{K_{i,s}(I - \Lambda_{i,s})r_{i,s}r_{i,s}^T(I - \Lambda_{i,s})K_{i,s}^T\} \\
&\quad + \mathbb{E}\{(I - K_{i,s}C_{i,s})\tilde{x}_{i,s|s-1}r_{i,s}^T(I - \Lambda_{i,s})K_{i,s}^T\} \\
&\quad + \mathbb{E}\{K_{i,s}(I - \Lambda_{i,s})r_{i,s}\tilde{x}_{i,s|s-1}^T(I - K_{i,s}C_{i,s})^T\} \\
&\quad - \mathbb{E}\{K_{i,s}\nu_{i,s}r_{i,s}^T(I - \Lambda_{i,s})K_{i,s}^T\} \\
&\quad - \mathbb{E}\{K_{i,s}(I - \Lambda_{i,s})r_{i,s}\nu_{i,s}^T K_{i,s}^T\}.
\end{aligned} \tag{22}$$

Remark 3: The accurate expressions of the one-step prediction error covariance and the filtering error covariance have been parameterized in (20) and (22). Nevertheless, the simultaneous presence of the component-wise DETT mechanism, uncertain dynamical bias and nonlinearities renders it really troublesome to calculate the accurate covariances. In this regard, an alternative yet effective approach is to seek upper bounds on the error covariance matrices as the performance criteria. It is worth mentioning that, compared with the existing literature with traditional static event-triggered mechanism, the introduction of the auxiliary variable $\xi_{i,l,s}$ would contribute to additional complexity in handling the term $\mathbb{E}\{r_{i,s}r_{i,s}^T\}$ and the cross terms related to $r_{i,s}$ since the auxiliary variable is not accessible from the filter end. On the other hand, by introducing a diagonal indicator matrix $\Lambda_{i,s}$ (with diagonal elements being 1 or 0), the effects of the component-wise

event-triggered transmissions have been clearly reflected in the dynamics of filtering error covariance.

To deal with the complexity induced by the component-wise DETT mechanism, the following proposition provides a feasible approach to establishing an upper bound for the term $\mathbb{E}\{r_{i,s}r_{i,s}^T\}$.

Proposition 1: For the i -th node, let $0 < \delta_{i,l} < 1$ and $\rho_{i,l} > \frac{1}{\delta_{i,l}}$ ($l = 1, 2, \dots, m$). The solution to the following recursion

$$\bar{\Xi}_{i,l,s} = \delta_{i,l}\bar{\Xi}_{i,l,s-1} + \pi_{i,l} \tag{23}$$

with initial condition $\bar{\Xi}_{i,l,0} = \xi_{i,l,0} \geq 0$ is always an upper bound on the auxiliary variable $\xi_{i,l,s}$, i.e., $0 \leq \xi_{i,l,s} \leq \bar{\Xi}_{i,l,s}$. In addition, the following inequality

$$r_{i,s}r_{i,s}^T \leq v_{i,s}I \tag{24}$$

holds, where

$$v_{i,s} \triangleq \sum_{l=1}^m \left(\pi_{i,l} + \frac{1}{\rho_{i,l}} \bar{\Xi}_{i,l,s} \right).$$

Proof: Following the similar line of [7], it is clear that, if $0 < \delta_{i,l} < 1$ and $\rho_{i,l} > \frac{1}{\delta_{i,l}}$ ($l = 1, 2, \dots, m$) hold, one has $\xi_{i,l,s} \geq 0$.

Next, we will prove $\xi_{i,l,s} \leq \bar{\Xi}_{i,l,s}$ by using the mathematical induction. For $s = 0$, the result is obviously true. Assume that $\xi_{i,l,s-1} \leq \bar{\Xi}_{i,l,s-1}$ holds. Then, at time instant s , noting that $r_{i,l,s-1}^2 \geq 0$, it follows from (7) that

$$\begin{aligned}
\xi_{i,l,s} &= \delta_{i,l}\xi_{i,l,s-1} - r_{i,l,s-1}^2 + \pi_{i,l} \\
&\leq \delta_{i,l}\bar{\Xi}_{i,l,s-1} + \pi_{i,l} = \bar{\Xi}_{i,l,s}
\end{aligned} \tag{25}$$

which ends the proof of the first part.

In what follows, we move on to the proof of the second part. Recalling the property of the DETT mechanism (6)-(8) as well as the inequality (25), it is not difficult to see that

$$r_{i,l,s}^2 \leq \pi_{i,l} + \frac{1}{\rho_{i,l}}\xi_{i,l,s} \leq \pi_{i,l} + \frac{1}{\rho_{i,l}}\bar{\Xi}_{i,l,s}. \tag{26}$$

Then, we can obtain that

$$r_{i,s}r_{i,s}^T \leq \|r_{i,s}\|^2 I = \sum_{l=1}^m r_{i,l,s}^2 I \leq \sum_{l=1}^m \left(\pi_{i,l} + \frac{1}{\rho_{i,l}}\bar{\Xi}_{i,l,s} \right) I. \tag{27}$$

The proof is now complete. \blacksquare

For convenience of presentation, define

$$F_s = \begin{bmatrix} \bar{F}_s & 0 \\ 0 & 0 \end{bmatrix}, \quad \bar{I} = \begin{bmatrix} I & 0 \\ 0 & 0 \end{bmatrix}.$$

Based on the results in Proposition 1 as well as the parameterized one-step prediction error covariance and filtering error covariance in (20) and (22), we are going to establish their corresponding upper bounds in the following proposition.

Proposition 2: For the i -th node, given positive scalars $\alpha_{i,j}$ ($j = 1, 2, \dots, 5$) and $\beta_{i,j}$ ($j = 1, 2$), the solutions to

the following two difference equations

$$\begin{aligned}
 & \mathcal{P}_{i,s|s-1} \\
 &= (1 + \alpha_{i,3} + \alpha_{i,4})A_{i,s-1}\mathcal{P}_{i,s-1}A_{i,s-1}^T \\
 &+ \tau_i \text{tr}\{(1 + \alpha_{i,1})\mathcal{P}_{i,s-1} + (1 + \alpha_{i,1}^{-1})\hat{x}_{i,s-1}\hat{x}_{i,s-1}^T\}I \\
 &+ (1 + \alpha_{i,3}^{-1} + \alpha_{i,5})\left\{(1 + \alpha_{i,2})\kappa_s^2 \text{tr}\{\bar{I}\mathcal{P}_{i,s-1}\bar{I}\}I \right. \\
 &+ (1 + \alpha_{i,2}^{-1})F_{s-1}\mathcal{P}_{i,s-1}F_{s-1}^T \left. \right\} + Q_{i,s-1} \\
 &+ (1 + \alpha_{i,4}^{-1} + \alpha_{i,5}^{-1})N \sum_{j=1}^N d_{ij}^2 \Gamma \mathcal{P}_{j,s-1} \Gamma
 \end{aligned} \tag{28}$$

and

$$\begin{aligned}
 & \mathcal{P}_{i,s} \\
 &= (1 + \beta_{i,1})(I - K_{i,s}C_{i,s})\mathcal{P}_{i,s|s-1}(I - K_{i,s}C_{i,s})^T \\
 &+ (1 + \beta_{i,2})K_{i,s}R_{i,s}K_{i,s}^T \\
 &+ (1 + \beta_{i,1}^{-1} + \beta_{i,2}^{-1})v_{i,s}K_{i,s}(I - \Lambda_{i,s})K_{i,s}^T
 \end{aligned} \tag{29}$$

with initial conditions $\mathcal{P}_{i,0} = P_{i,0} > 0$ represent, respectively, the upper bounds on the one-step prediction error covariance and filtering error covariance, i.e.,

$$P_{i,s|s-1} \leq \mathcal{P}_{i,s|s-1}, \quad P_{i,s} \leq \mathcal{P}_{i,s}.$$

Proof: To begin with, let us analyze the unknown terms on the right-hand side of (20) one by one. Concentrating on the second term on the right-hand side of (20) and recalling the elementary inequality $xy^T + yx^T \leq \alpha_{i,1}xx^T + \alpha_{i,1}^{-1}yy^T$ for $x, y \in \mathbb{R}^n$ and any positive scalar $\alpha_{i,1}$, it is not difficult to obtain that

$$\begin{aligned}
 & \mathbb{E}\{\Delta A_{i,s-1}x_{i,s-1}x_{i,s-1}^T \Delta A_{i,s-1}^T\} \\
 & \leq \tau_i \text{tr}\{(1 + \alpha_{i,1})P_{i,s-1} + (1 + \alpha_{i,1}^{-1})\hat{x}_{i,s-1}\hat{x}_{i,s-1}^T\}I.
 \end{aligned} \tag{30}$$

In light of (3), the third term on the right-hand side of (20) can be rearranged as follows:

$$\begin{aligned}
 & \mathbb{E}\{\tilde{f}_{s-1}(x_{i,s-1})\tilde{f}_{s-1}(x_{i,s-1})^T\} \\
 & \leq \mathbb{E}\{(1 + \alpha_{i,2})\|f_{s-1}(x_{i,s-1}) \\
 & \quad - f_{s-1}(\hat{x}_{i,s-1}) - F_{s-1}\tilde{x}_{i,s-1}\|^2 I \\
 & \quad + (1 + \alpha_{i,2}^{-1})F_{s-1}\tilde{x}_{i,s-1}\tilde{x}_{i,s-1}^T F_{s-1}^T\} \\
 & \leq \mathbb{E}\{(1 + \alpha_{i,2})\kappa_s^2 \|\bar{I}\tilde{x}_{i,s-1}\|^2 I \\
 & \quad + (1 + \alpha_{i,2}^{-1})F_{s-1}\tilde{x}_{i,s-1}\tilde{x}_{i,s-1}^T F_{s-1}^T\} \\
 & = (1 + \alpha_{i,2})\kappa_s^2 \text{tr}\{\bar{I}P_{i,s-1}\bar{I}\}I \\
 & \quad + (1 + \alpha_{i,2}^{-1})F_{s-1}P_{i,s-1}F_{s-1}^T
 \end{aligned} \tag{31}$$

where $\alpha_{i,2}$ is a positive scalar.

Now, we focus on the fifth term on the right-hand side of (20). Based on the aforementioned elementary inequality, it is straightforward to see that

$$\begin{aligned}
 & \mathbb{E}\left\{\left(\sum_{j=1}^N d_{ij}\Gamma\tilde{x}_{j,s-1}\right)\left(\sum_{j=1}^N d_{ij}\Gamma\tilde{x}_{j,s-1}\right)^T\right\} \\
 & \leq N \sum_{j=1}^N d_{ij}^2 \Gamma \mathcal{P}_{j,s-1} \Gamma.
 \end{aligned} \tag{32}$$

Subsequently, let us move on to deal with the cross terms in (20). Selecting the positive scalars $\alpha_{i,3}$, $\alpha_{i,4}$ and $\alpha_{i,5}$, one has

$$\begin{aligned}
 & \mathbb{E}\{A_{i,s-1}\tilde{x}_{i,s-1}\tilde{f}_{s-1}(x_{i,s-1})^T\} \\
 & + \mathbb{E}\{\tilde{f}_{s-1}(x_{i,s-1})\tilde{x}_{i,s-1}^T A_{i,s-1}^T\} \\
 & \leq \alpha_{i,3}A_{i,s-1}P_{i,s-1}A_{i,s-1}^T \\
 & \quad + \alpha_{i,3}^{-1}\mathbb{E}\{\tilde{f}_{s-1}(x_{i,s-1})\tilde{f}_{s-1}(x_{i,s-1})^T\},
 \end{aligned} \tag{33}$$

$$\begin{aligned}
 & \mathbb{E}\{A_{i,s-1}\tilde{x}_{i,s-1}\left(\sum_{j=1}^N d_{ij}\Gamma\tilde{x}_{j,s-1}\right)^T\} \\
 & + \mathbb{E}\left\{\left(\sum_{j=1}^N d_{ij}\Gamma\tilde{x}_{j,s-1}\right)\tilde{x}_{i,s-1}^T A_{i,s-1}^T\right\} \\
 & \leq \alpha_{i,4}A_{i,s-1}P_{i,s-1}A_{i,s-1}^T \\
 & \quad + \alpha_{i,4}^{-1}\mathbb{E}\left\{\left(\sum_{j=1}^N d_{ij}\Gamma\tilde{x}_{j,s-1}\right)\left(\sum_{j=1}^N d_{ij}\Gamma\tilde{x}_{j,s-1}\right)^T\right\},
 \end{aligned} \tag{34}$$

and

$$\begin{aligned}
 & \mathbb{E}\{\tilde{f}_{s-1}(x_{i,s-1})\left(\sum_{j=1}^N d_{ij}\Gamma\tilde{x}_{j,s-1}\right)^T\} \\
 & + \mathbb{E}\left\{\left(\sum_{j=1}^N d_{ij}\Gamma\tilde{x}_{j,s-1}\right)\tilde{f}_{s-1}(\tilde{x}_{i,s-1})^T\right\} \\
 & \leq \alpha_{i,5}\mathbb{E}\{\tilde{f}_{s-1}(x_{i,s-1})\tilde{f}_{s-1}(\tilde{x}_{i,s-1})^T\} \\
 & \quad + \alpha_{i,5}^{-1}\mathbb{E}\left\{\left(\sum_{j=1}^N d_{ij}\Gamma\tilde{x}_{j,s-1}\right)\left(\sum_{j=1}^N d_{ij}\Gamma\tilde{x}_{j,s-1}\right)^T\right\}.
 \end{aligned} \tag{35}$$

Substituting (30)-(35) into (20) results in

$$\begin{aligned}
 & P_{i,s|s-1} \\
 & \leq (1 + \alpha_{i,3} + \alpha_{i,4})A_{i,s-1}P_{i,s-1}A_{i,s-1}^T \\
 & + \tau_i \text{tr}\{(1 + \alpha_{i,1})P_{i,s-1} + (1 + \alpha_{i,1}^{-1})\hat{x}_{i,s-1}\hat{x}_{i,s-1}^T\}I \\
 & + (1 + \alpha_{i,3}^{-1} + \alpha_{i,5})\left\{(1 + \alpha_{i,2})\kappa_s^2 \text{tr}\{\bar{I}P_{i,s-1}\bar{I}\}I \right. \\
 & + (1 + \alpha_{i,2}^{-1})F_{s-1}P_{i,s-1}F_{s-1}^T \left. \right\} + Q_{i,s-1} \\
 & + (1 + \alpha_{i,4}^{-1} + \alpha_{i,5}^{-1})N \sum_{j=1}^N d_{ij}^2 \Gamma \mathcal{P}_{j,s-1} \Gamma.
 \end{aligned} \tag{36}$$

Next, let us consider the cross terms on the right-hand side of (22). Similar to the above derivations, the cross terms satisfy the following relationships:

$$\begin{aligned}
 & \mathbb{E}\{(I - K_{i,s}C_{i,s})\tilde{x}_{i,s|s-1}r_{i,s}^T(I - \Lambda_{i,s})K_{i,s}^T\} \\
 & + \mathbb{E}\{K_{i,s}(I - \Lambda_{i,s})r_{i,s}\tilde{x}_{i,s|s-1}^T(I - K_{i,s}C_{i,s})^T\} \\
 & \leq \beta_{i,1}(I - K_{i,s}C_{i,s})P_{i,s|s-1}(I - K_{i,s}C_{i,s})^T \\
 & \quad + \beta_{i,1}^{-1}\mathbb{E}\{K_{i,s}(I - \Lambda_{i,s})r_{i,s}r_{i,s}^T(I - \Lambda_{i,s})K_{i,s}^T\}
 \end{aligned} \tag{37}$$

and

$$\begin{aligned}
 & - \mathbb{E}\{K_{i,s}v_{i,s}r_{i,s}^T(I - \Lambda_{i,s})K_{i,s}^T\} \\
 & - \mathbb{E}\{K_{i,s}(I - \Lambda_{i,s})r_{i,s}v_{i,s}^T K_{i,s}^T\} \\
 & \leq \beta_{i,2}K_{i,s}R_{i,s}K_{i,s}^T \\
 & \quad + \beta_{i,2}^{-1}\mathbb{E}\{K_{i,s}(I - \Lambda_{i,s})r_{i,s}r_{i,s}^T(I - \Lambda_{i,s})K_{i,s}^T\}.
 \end{aligned} \tag{38}$$

Recalling (24) and the fact that $(I - \Lambda_{i,s})(I - \Lambda_{i,s}) = (I - \Lambda_{i,s})$, one has

$$\begin{aligned} & \mathbb{E}\{K_{i,s}(I - \Lambda_{i,s})r_{i,s}r_{i,s}^T(I - \Lambda_{i,s})K_{i,s}^T\} \\ & \leq v_{i,s}K_{i,s}(I - \Lambda_{i,s})K_{i,s}^T. \end{aligned} \quad (39)$$

Then, substituting (37)-(39) into (22) leads to

$$\begin{aligned} & P_{i,s} \\ & \leq (1 + \beta_{i,1})(I - K_{i,s}C_{i,s})P_{i,s|s-1}(I - K_{i,s}C_{i,s})^T \\ & \quad + (1 + \beta_{i,2})K_{i,s}R_{i,s}K_{i,s}^T \\ & \quad + (1 + \beta_{i,1}^{-1} + \beta_{i,2}^{-1})v_{i,s}K_{i,s}(I - \Lambda_{i,s})K_{i,s}^T. \end{aligned} \quad (40)$$

Based on the relationships (36) and (40) as well as the mathematical induction method, it is not difficult to verify that $\mathcal{P}_{i,s|s-1}$ and $\mathcal{P}_{i,s}$ are, respectively, the upper bounds on the one-step prediction error covariance and the filtering error covariance. The proof is now complete. ■

In view of Proposition 2, the weighted matrix $\mathcal{P}_{i,s|s-1}$ is parameterized by the one-step prediction error covariance. Next, we further determine the form of another weighted matrix $\mathcal{R}_{i,s}$ by considering the effect of the component-wise DETT mechanism. From (10), it is clear that

$$\begin{aligned} & \bar{y}_{i,s} - C_{i,s}x_{i,s} \\ & = y_{i,s} - (I - \Lambda_{i,s})r_{i,s} - C_{i,s}x_{i,s} \\ & = C_{i,s}x_{i,s} + \nu_{i,s} - (I - \Lambda_{i,s})r_{i,s} - C_{i,s}x_{i,s} \\ & = \nu_{i,s} - (I - \Lambda_{i,s})r_{i,s}. \end{aligned} \quad (41)$$

Then, one has

$$\begin{aligned} & \mathbb{E}\{(\bar{y}_{i,s} - C_{i,s}x_{i,s})(\bar{y}_{i,s} - C_{i,s}x_{i,s})^T\} \\ & = \mathbb{E}\{(\nu_{i,s} - (I - \Lambda_{i,s})r_{i,s})(\nu_{i,s} - (I - \Lambda_{i,s})r_{i,s})^T\} \\ & \leq (1 + \beta_{i,2})R_{i,s} + (1 + \beta_{i,2}^{-1})v_{i,s}(I - \Lambda_{i,s}) \\ & \triangleq \mathcal{R}_{i,s}, \end{aligned} \quad (42)$$

which can be regarded as an upper bound on the equivalent measurement noise covariance.

By resorting to the weighted matrices $\mathcal{P}_{i,s|s-1}$ and $\mathcal{R}_{i,s}$ determined in (28) and (42), we are able to parameterize the explicit expression of the CBPI (16). Accordingly, motivated by [16], the MCF scheme is designed as follows.

Bearing in mind the fact that the design purpose of filter (i.e., gain $K_{i,s}$) is to maximize the CBPI (16), we take partial derivative with respect to $x_{i,s}$ and let

$$\frac{\partial J(x_{i,s})}{\partial x_{i,s}} = 0.$$

Then, it is obtained that

$$\begin{aligned} & -\frac{1}{\chi^2}G_\chi \left(\|x_{i,s} - \hat{x}_{i,s|s-1}\|_{\mathcal{P}_{i,s|s-1}^{-1}} \right) \\ & \quad \times \mathcal{P}_{i,s|s-1}^{-1}(x_{i,s} - \hat{x}_{i,s|s-1}) \\ & + \frac{1}{\chi^2}G_\chi \left(\|\bar{y}_{i,s} - C_{i,s}x_{i,s}\|_{\mathcal{R}_{i,s}^{-1}} \right) \\ & \quad \times C_{i,s}^T \mathcal{R}_{i,s}^{-1}(\bar{y}_{i,s} - C_{i,s}x_{i,s}) = 0. \end{aligned} \quad (43)$$

For ease of analysis, let us denote

$$U_{i,s} = \frac{G_\chi \left(\|\bar{y}_{i,s} - C_{i,s}x_{i,s}\|_{\mathcal{R}_{i,s}^{-1}} \right)}{G_\chi \left(\|x_{i,s} - \hat{x}_{i,s|s-1}\|_{\mathcal{P}_{i,s|s-1}^{-1}} \right)}. \quad (44)$$

Then, (43) can be rewritten by

$$\mathcal{P}_{i,s|s-1}^{-1}(x_{i,s} - \hat{x}_{i,s|s-1}) = U_{i,s}C_{i,s}^T \mathcal{R}_{i,s}^{-1}(\bar{y}_{i,s} - C_{i,s}x_{i,s}), \quad (45)$$

which, by adding the zero term

$$U_{i,s}C_{i,s}^T \mathcal{R}_{i,s}^{-1}C_{i,s}\hat{x}_{i,s|s-1} - U_{i,s}C_{i,s}^T \mathcal{R}_{i,s}^{-1}C_{i,s}\hat{x}_{i,s|s-1},$$

can be further reformulated as

$$\begin{aligned} & (\mathcal{P}_{i,s|s-1}^{-1} + U_{i,s}C_{i,s}^T \mathcal{R}_{i,s}^{-1}C_{i,s})x_{i,s} \\ & = (\mathcal{P}_{i,s|s-1}^{-1} + U_{i,s}C_{i,s}^T \mathcal{R}_{i,s}^{-1}C_{i,s})\hat{x}_{i,s|s-1} \\ & \quad + U_{i,s}C_{i,s}^T \mathcal{R}_{i,s}^{-1}(\bar{y}_{i,s} - C_{i,s}\hat{x}_{i,s|s-1}). \end{aligned} \quad (46)$$

Clearly, we can obtain that

$$\hat{x}_{i,s} = \hat{x}_{i,s|s-1} + K_{i,s}(\bar{y}_{i,s} - C_{i,s}\hat{x}_{i,s|s-1}) \quad (47)$$

where

$$K_{i,s} = (\mathcal{P}_{i,s|s-1}^{-1} + U_{i,s}C_{i,s}^T \mathcal{R}_{i,s}^{-1}C_{i,s})^{-1}U_{i,s}C_{i,s}^T \mathcal{R}_{i,s}^{-1}. \quad (48)$$

It should be pointed out that $x_{i,s}$ is also involved in the expression of $U_{i,s}$ (and hence in $K_{i,s}$), which means that it would be really tricky to derive an analytic form of the state estimate $\hat{x}_{i,s}$ via (47). To this end, the fixed-point iterative algorithm is introduced in [5] with the iterative initial state being the one-step prediction. In this paper, following the similar line of [22], [45], the fixed-point algorithm with only one iteration is employed to guarantee the calculation efficiency. Specifically, the one-step prediction $\hat{x}_{i,s|s-1}$ is utilized to approximate $x_{i,s}$ in both the numerator and denominator on the right-hand side of (44), which implies that

$$\hat{U}_{i,s} = G_\chi \left(\|\bar{y}_{i,s} - C_{i,s}\hat{x}_{i,s|s-1}\|_{\mathcal{R}_{i,s}^{-1}} \right). \quad (49)$$

Then, the filter gain matrix can be calculated by

$$K_{i,s} = (\mathcal{P}_{i,s|s-1}^{-1} + \hat{U}_{i,s}C_{i,s}^T \mathcal{R}_{i,s}^{-1}C_{i,s})^{-1}\hat{U}_{i,s}C_{i,s}^T \mathcal{R}_{i,s}^{-1}. \quad (50)$$

To facilitate the future implementation, the developed MCF algorithm under the component-wise DETT mechanism is summarized in Algorithm 1.

Remark 4: It should be emphasized that, for the traditional Kalman-like filtering algorithms based on the variance-constrained approach, the filter gain matrix is designed to minimize the trace of the filtering error covariance or its upper bound by letting

$$\frac{\partial \text{tr}(\mathcal{P}_{i,s})}{\partial K_{i,s}} = 0.$$

In this case, recalling the expression of $\mathcal{P}_{i,s}$ in (29), one has

$$\begin{aligned} & (1 + \beta_{i,1})(-2\mathcal{P}_{i,s|s-1}C_{i,s}^T + 2K_{i,s}C_{i,s}\mathcal{P}_{i,s|s-1}C_{i,s}^T) \\ & \quad + 2(1 + \beta_{i,2})K_{i,s}R_{i,s} \\ & \quad + 2(1 + \beta_{i,1}^{-1} + \beta_{i,2}^{-1})v_{i,s}K_{i,s}(I - \Lambda_{i,s}) = 0. \end{aligned}$$

Then, the filter gain matrix for the i -th node can be obtained as follows:

$$\begin{aligned} K_{i,s} = & (1 + \beta_{i,1})\mathcal{P}_{i,s|s-1}C_{i,s}^T \left[(1 + \beta_{i,1})C_{i,s}\mathcal{P}_{i,s|s-1}C_{i,s}^T \right. \\ & \quad + (1 + \beta_{i,2})R_{i,s} \\ & \quad \left. + (1 + \beta_{i,1}^{-1} + \beta_{i,2}^{-1})v_{i,s}(I - \Lambda_{i,s}) \right]^{-1}. \end{aligned} \quad (51)$$

Algorithm 1 MCF algorithm under the component-wise DETT mechanism.

Step 1. Algorithm initialization

For each node i ($i = 1, 2, \dots, N$), set the initial estimates $\hat{x}_{i,0}$ and initial covariance matrices $\mathcal{P}_{i,0}$. Choose the positive scalars $\alpha_{i,j}$ ($j = 1, 2, \dots, 5$) and $\beta_{i,j}$ ($j = 1, 2$), and select the proper kernel bandwidth χ . The maximum time instant is set to be K .

Step 2. One-step prediction

Calculate the one-step prediction $\hat{x}_{i,s|s-1}$ and the upper bound $\mathcal{P}_{i,s|s-1}$ on the prediction error covariance based on (14) and (28), respectively.

Step 3. State estimate update

- Collect the available measurements $\bar{y}_{i,s}$ under scheduling of the component-wise DETT scheme (6) and (7).
- Calculate the filter gain matrix $K_{i,s}$ based on (42), (49) and (50).
- Update the state estimate $\hat{x}_{i,s}$ and the upper bound $\mathcal{P}_{i,s}$ on the filtering error covariance based on (15) and (29), respectively.

Step 4. If $s < K$, then set $s = s + 1$ and go to **Step 2**; otherwise stop the recursion.

It is clear that the above derivations only make use of the second-order moment information of the filtering error, and hence the resulted filter would be inapplicable to the scenarios with non-Gaussian noises. Frankly speaking, such a problem greatly motivates this current investigation.

Remark 5: It should also be mentioned that the number of required floating point operations has been widely utilized to approximately evaluate the computational complexity of a given algorithm. Following the similar line of [5], we can conclude that the computational complexity of the proposed MCF algorithm (composed of (14), (15), (28), (29), (42), (49) and (50)) is $C_1 = 12n^3 + (14 + 4N)n^2 + 12mn^2 + 4m^2 + 6nm^2 + (3N + 4)n + 6mn + 9m + \Delta_f + N + 22 + O(m^3) + 2O(n^3)$ (where $n = \bar{n} + k$ and Δ_f denote, respectively, the state dimension of the augmented system (5) and the computational complexity determined by the specific form of nonlinear function $f_s(\cdot)$), and that of the variance-constrained filtering algorithm (comprised by (14), (15), (28), (29) and (51)) is $C_2 = 12n^3 + (13 + 4N)n^2 + 12mn^2 + 3m^2 + 6nm^2 + (3N + 4)n + 2mn + 8m + \Delta_f + N + 23 + O(m^3)$. In summary, the proposed algorithm has moderate computational complexity in comparison with the traditional variance-constrained one, while guaranteeing a substantial performance improvement under the non-Gaussian environment.

Remark 6: So far, considerable research enthusiasms have been devoted to the state estimation or filtering problem for complex networks with various network-induced phenomena, and a rich body of elegant results has been reported in recent years. Compared with the existing literature, this paper may shed some new insights from the following three aspects: 1) the problem under investigation is new in the sense that it takes

into simultaneous account the non-Gaussian noises, the system nonlinearity, the uncertain dynamical bias and the component-wise DETT protocol; 2) the maximum correntropy criterion is adopted with a novel performance index which embraces the effects of uncertain dynamical bias and component-wise DETT protocol; and 3) the proposed filtering algorithm inherits the recursive form of the Kalman-like filters and exhibits appealing robustness against the non-Gaussian noises under the component-wise DETT mechanism, which is therefore suitable for the online implementations in practice.

IV. SIMULATION VALIDATIONS

In this section, an illustrative example is presented to demonstrate the effectiveness of the developed filtering methodology. Consider the complex network (1)-(4) with six nodes as well as the following parameters:

$$\begin{aligned} \bar{A}_{1,s} &= \begin{bmatrix} 0.85 + 0.1 \sin(s) & 0.6 \\ -0.2 & 0.85 \end{bmatrix}, B_{1,s} = \begin{bmatrix} 0.4 & 0.5 \\ 0.4 & -0.25 \end{bmatrix} \\ \bar{A}_{2,s} &= \begin{bmatrix} 0.7 & -0.05 + 0.1 \sin(s) \\ -0.3 & 0.7 \end{bmatrix}, B_{2,s} = \begin{bmatrix} 0.8 & 0.4 \\ 0.3 & -0.2 \end{bmatrix} \\ \bar{A}_{3,s} &= \begin{bmatrix} 0.8 & 0.4 \\ -0.4 & 0.65 + 0.2 \cos(s) \end{bmatrix}, B_{3,s} = \begin{bmatrix} 0.7 & -0.1 \\ 0.2 & 0.3 \end{bmatrix}, \\ \bar{A}_{4,s} &= \begin{bmatrix} 0.8 + 0.1 \cos(s) & 0.4 \\ -0.2 & 0.75 \end{bmatrix}, B_{4,s} = \begin{bmatrix} 0.6 & -0.1 \\ 0.2 & 0.2 \end{bmatrix}, \\ \bar{A}_{5,s} &= \begin{bmatrix} 0.9 & 0.1 \\ -0.1 + 0.1 \sin(s) & 0.7 \end{bmatrix}, B_{5,s} = \begin{bmatrix} 0.5 & -0.1 \\ 0.1 & 0.3 \end{bmatrix}, \\ \bar{A}_{6,s} &= \begin{bmatrix} 0.85 & -0.05 + 0.1 \cos(s) \\ -0.1 & 0.75 \end{bmatrix}, B_{6,s} = \begin{bmatrix} 0.6 & -0.1 \\ 0.1 & 0.2 \end{bmatrix}, \\ G_{1,s} &= \begin{bmatrix} 0.5 & 0.4 \\ -0.3 & 0.6 \end{bmatrix}, G_{2,s} = \begin{bmatrix} 0.6 & 0.4 \\ 0.4 & 0.3 \end{bmatrix}, \\ G_{3,s} &= \begin{bmatrix} 0.6 & 0.2 \\ 0.3 & 0.4 \end{bmatrix}, G_{4,s} = \begin{bmatrix} 0.3 & 0.2 \\ 0.1 & 0.3 \end{bmatrix}, G_{5,s} = \begin{bmatrix} 0.5 & 0.2 \\ 0.1 & 0.2 \end{bmatrix}, \\ G_{6,s} &= \begin{bmatrix} 0.4 & -0.1 \\ 0.1 & 0.2 \end{bmatrix}, \bar{C}_{1,s} = \begin{bmatrix} 1.2 & 0 \\ 0 & 1.3 \end{bmatrix}, \\ \bar{C}_{2,s} &= \begin{bmatrix} 1.8 & 0 \\ 0 & 1.2 \end{bmatrix}, \bar{C}_{3,s} = \begin{bmatrix} 1.5 & 0 \\ 0 & 1.4 \end{bmatrix}, \bar{C}_{4,s} = \begin{bmatrix} 1.6 & 0 \\ 0 & 1.4 \end{bmatrix}, \\ \bar{C}_{5,s} &= \begin{bmatrix} 1.9 & 0 \\ 0 & 1.5 \end{bmatrix}, \bar{C}_{6,s} = \begin{bmatrix} 1.5 & 0 \\ 0 & 1.2 \end{bmatrix}, \bar{\Gamma} = \begin{bmatrix} 0.1 & 0 \\ 0 & 0.1 \end{bmatrix}. \end{aligned}$$

The nonlinear function is described by

$$\bar{f}_s(\bar{x}_{i,s}) = \begin{bmatrix} 0.1 & -0.01 + 0.05 \cos(s) \\ 0.01 & 0.15 \end{bmatrix} \bar{x}_{i,s} + \begin{bmatrix} 0.08 \sin(\bar{x}_{i,s}^1) \\ 0.1 \sin(\bar{x}_{i,s}^2) \end{bmatrix}$$

where $\bar{x}_{i,s}^l$ ($l = 1, 2$) denotes the l th component of the state $\bar{x}_{i,s}$. Other modeling parameters are selected as $\tau_i = 0.01$, $d_{ij} = 0.2$ ($i \neq j$) and $d_{ii} = -1$ for $i = 1, 2, \dots, 6$.

In the simulation, the initial conditions are chosen as $\bar{x}_{1,0} = [0 \ 0]^T$, $\bar{x}_{2,0} = [-1 \ 1]^T$, $\bar{x}_{3,0} = [2 \ -1]^T$, $\bar{x}_{4,0} = [-1 \ 2]^T$, $\bar{x}_{5,0} = [-2 \ -1]^T$, $\bar{x}_{6,0} = [0 \ -1]^T$, $z_{1,0} = [-1 \ -2]^T$, $z_{2,0} = [2 \ 3]^T$, $z_{3,0} = [2 \ -4]^T$, $z_{4,0} = [2 \ -3]^T$, $z_{5,0} = [-1 \ 2]^T$, $z_{6,0} = [1 \ 2]^T$ and $P_{i,0} = 2I$ ($i = 1, 2, \dots, 6$). For the component-wise DETT protocol, we let $\rho_{i,l} = 4$, $\delta_{i,l} = 0.9$, $\xi_{i,l,0} = 1$ for $i = 1, 2, \dots, 6$ and $l = 1, 2$. Moreover, $\pi_{1,1} = 1$, $\pi_{1,2} = 2$, $\pi_{2,1} = 2$, $\pi_{2,2} = 1$, $\pi_{3,1} = 1$, $\pi_{3,2} = 1$, $\pi_{4,1} = 1.5$, $\pi_{4,2} = 2$, $\pi_{5,1} = 1$, $\pi_{5,2} = 1.2$, $\pi_{6,1} = 1.2$

and $\pi_{6,2} = 1$. The root mean square error (RMSE) on state estimate is calculated over 500 Monte Carlo runs to facilitate the evaluation/comparison of filtering performance.

In what follows, two scenarios with different noise conditions are taken into account to compare the tracking performance of the proposed MCF algorithm (abbreviated as PMCF) and the variance-constrained filtering algorithm (denoted as VCF) in Remark 4.

Case 1: In this case, both the process noise and measurement noise are Gaussian mixture noises, i.e.,

$$\begin{aligned} \zeta_{i,s} &\sim 0.9\mathcal{N}(0, \text{diag}\{0.01, 0.01\}) + 0.1\mathcal{N}(0, \text{diag}\{1, 1\}), \\ \eta_{i,s} &\sim 0.9\mathcal{N}(0, \text{diag}\{0.001, 0.001\}) + 0.1\mathcal{N}(0, \text{diag}\{0.1, 0.1\}), \\ \nu_{i,s} &\sim 0.8\mathcal{N}(0, \text{diag}\{0.5, 0.5\}) + 0.2\mathcal{N}(0, \text{diag}\{500, 500\}). \end{aligned}$$

The simulation results are shown in Figs. 1-4. It can be observed from Figs. 1-3 that compared with the VCF method, our proposed algorithm can well track the true state trajectories of three representative nodes, i.e. nodes 1, 3 and 5. In Fig. 4, the triggering instants for the first node are displayed, which verifies that the sensor components have different triggering rates under the component-wise DETT protocol.

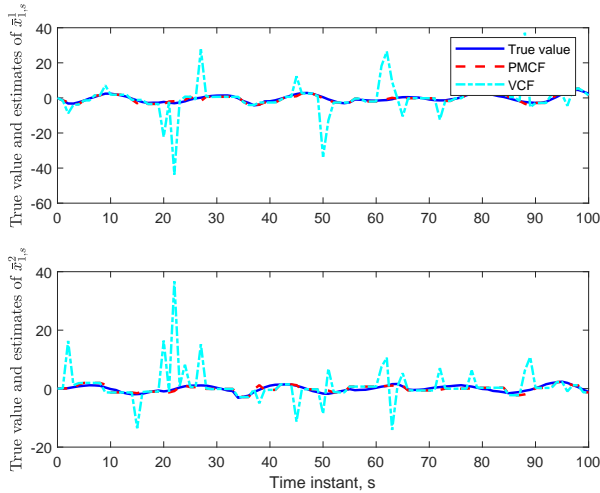


Fig. 1: True value and state estimates for node 1 in *Case 1*.

The average RMSEs on state estimate, calculated over the total simulation times, are summarized in Table I, where only the results for nodes 1, 3 and 5 are displayed for brevity. Obviously, in this case, the kernel bandwidth can be neither too large nor too small. When the kernel bandwidth becomes very larger, the performance of the proposed algorithm is similar to that of the VCF method. It should be pointed out that when the kernel bandwidth is set as $\chi = 0.08$ in this case, the proposed algorithm has a noticeable performance improvement over the traditional method based on variance-constrained strategy.

To examine the effect of the triggering parameters onto the filtering performance and resource consumption, the average transmission rate is defined as the mean of the transmission rate (the ratio of the triggering times to total simulation times) over 500 Monte Carlo runs. The simulation results in terms of average RMSEs and transmission rates are provided in Table II, from which we are able to conclude that a proper

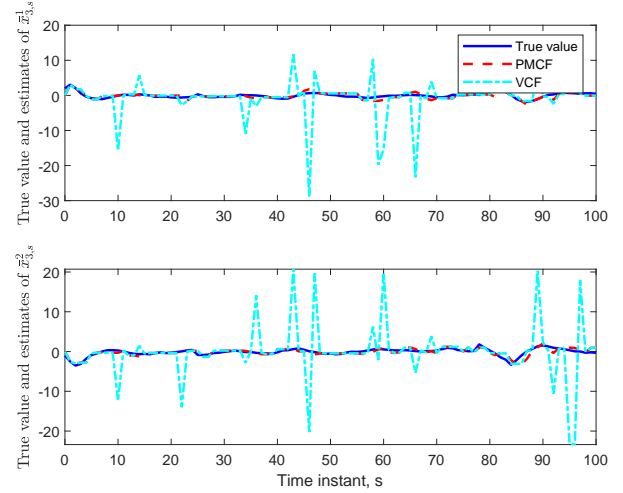


Fig. 2: True value and state estimates for node 3 in *Case 1*.

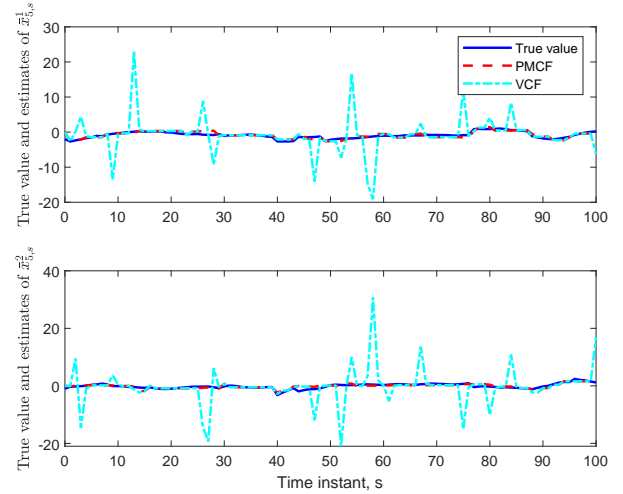


Fig. 3: True value and state estimates for node 5 in *Case 1*.

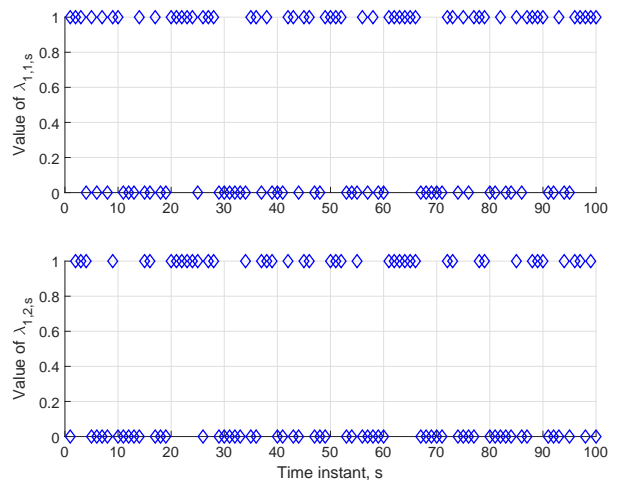


Fig. 4: Triggering instants for node 1 in *Case 1*.

TABLE I: Performance comparisons for different algorithms and parameters under Gaussian mixture noises.

	Node 1		Node 3		Node 5	
	RMSE ₁ ¹	RMSE ₁ ²	RMSE ₃ ¹	RMSE ₃ ²	RMSE ₅ ¹	RMSE ₅ ²
VCF	8.2005	7.4317	6.4892	6.9062	5.0592	6.1273
PMCF ($\chi = 0.01$)	1.6610	1.0168	0.8645	0.7956	0.8902	0.6973
PMCF ($\chi = 0.08$)	0.9584	0.8249	0.8456	0.7044	0.5898	0.5891
PMCF ($\chi = 0.5$)	4.2965	3.0973	4.3674	2.7436	2.0805	1.7407
PMCF ($\chi = 5$)	8.1172	7.3320	6.3757	6.7044	4.9704	5.8088
PMCF ($\chi = 10$)	8.1419	7.3618	6.4129	6.7666	4.9972	5.9080

TABLE II: Average RMSEs and transmission rates for different triggering parameters under Gaussian mixture noises.

		$\rho_{i,l} = 1.2$	$\rho_{i,l} = 4$	$\rho_{i,l} = \infty$
		Node 1	RMSE ₁ ¹	1.0465
	RMSE ₁ ²	0.9180	0.8249	0.7127
	TR ₁ ¹	49.65%	54.62%	66.37%
	TR ₁ ²	38.44%	41.60%	51.31%
Node 2	RMSE ₂ ¹	0.9543	0.8686	0.7283
	RMSE ₂ ²	0.8779	0.8010	0.6956
	TR ₂ ¹	38.59%	41.74%	52.57%
	TR ₂ ²	41.22%	45.87%	59.14%
Node 3	RMSE ₃ ¹	0.9020	0.8456	0.7735
	RMSE ₃ ²	0.7565	0.7044	0.6381
	TR ₃ ¹	45.58%	51.02%	63.31%
	TR ₃ ²	44.81%	50.21%	62.61%
Node 4	RMSE ₄ ¹	0.7556	0.6980	0.6368
	RMSE ₄ ²	0.7736	0.7137	0.6319
	TR ₄ ¹	41.58%	45.66%	56.33%
	TR ₄ ²	36.81%	39.88%	49.53%
Node 5	RMSE ₅ ¹	0.5928	0.5898	0.5611
	RMSE ₅ ²	0.6190	0.5891	0.5515
	TR ₅ ¹	42.33%	47.15%	59.74%
	TR ₅ ²	38.14%	42.67%	55.08%
Node 6	RMSE ₆ ¹	0.6299	0.5976	0.5639
	RMSE ₆ ²	0.6897	0.6666	0.6321
	TR ₆ ¹	38.31%	42.83%	55.62%
	TR ₆ ²	38.66%	43.79%	56.52%

triggering parameter can be chosen to cater for the practical engineering requirements. On the other hand, it is worth mentioning that when $\rho_{i,l} = \infty$, the component-wise DETT protocol reduces to its component-wise static counterpart. Clearly, the dynamic version has great potentials in relaxing the network communication burden.

Case 2: In this case, the process noise is of the Gaussian form and the measurement noise is shot noise, i.e.,

$$\begin{aligned} \zeta_{i,s} &\sim \mathcal{N}(0, \text{diag}\{0.01, 0.01\}), \\ \eta_{i,s} &\sim \mathcal{N}(0, \text{diag}\{0.001, 0.001\}), \\ \nu_{i,s} &\sim \mathcal{N}(0, \text{diag}\{0.5, 0.5\}) + \text{Shot noise}. \end{aligned}$$

For the first sensor component of node 1, one realization of the measurement noise at the time instants when shot noise occurs is depicted in Fig. 5. The simulation results are summarized in Figs. 6-12, which again verify the effectiveness

of the proposed filtering scheme in terms of dealing with the typical shot non-Gaussian noises and saving the limited network resources.

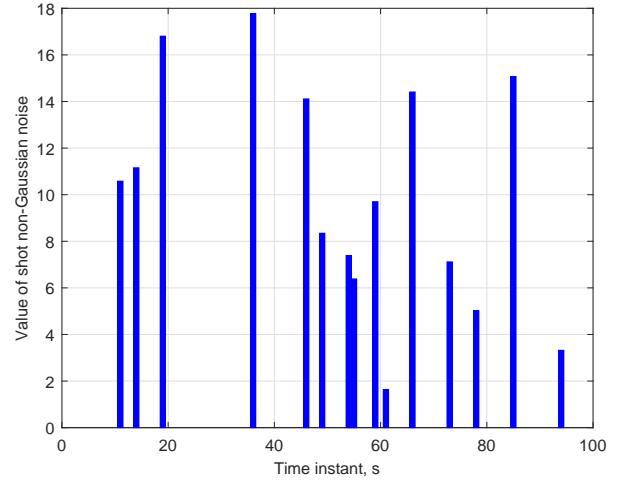


Fig. 5: One realization of the measurement noise at the time instants when shot noise occurs.

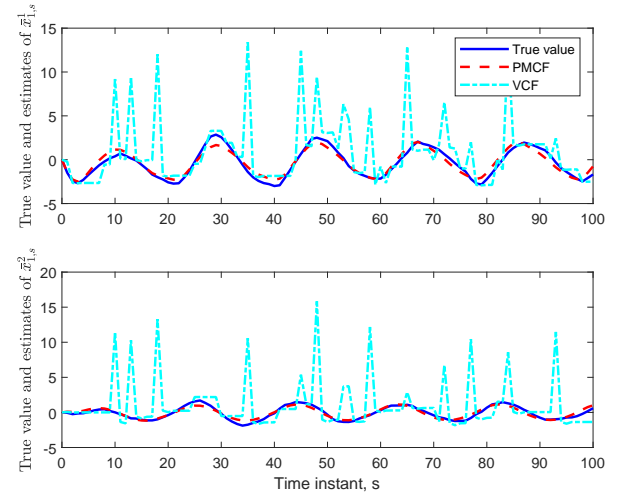


Fig. 6: True value and state estimates for node 1 in Case 2.

V. CONCLUSIONS

In this paper, the MCF problem has been dealt with for a class of nonlinear and non-Gaussian complex networks subject to uncertain dynamical bias under the DETT mechanism. To guarantee that each sensor component can determine its own triggering instant in an independent manner, the component-wise DETT protocol has been exploited to govern the process of data transmission. By resorting to the established upper bounds on the one-step prediction error covariance and the equivalent noise covariance, a novel CBPI has been parameterized to extenuate the influence of the non-Gaussian noises and the component-wise DETT protocol. Accordingly, a recursive filter has been designed based on the maximum

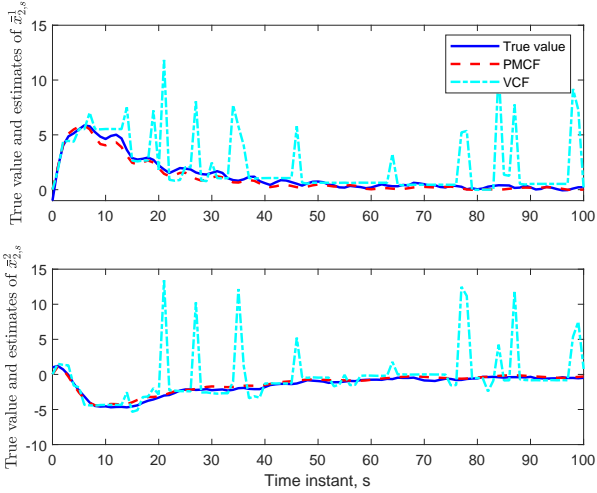


Fig. 7: True value and state estimates for node 2 in *Case 2*.

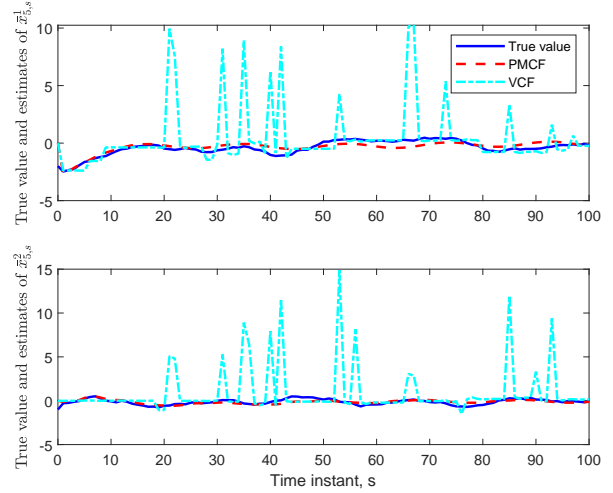


Fig. 10: True value and state estimates for node 5 in *Case 2*.

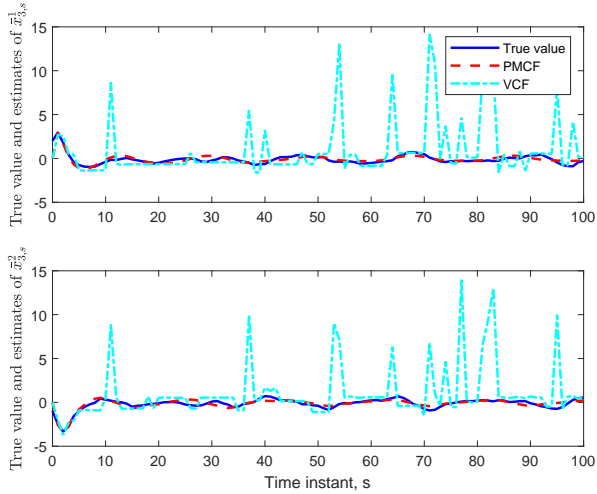


Fig. 8: True value and state estimates for node 3 in *Case 2*.

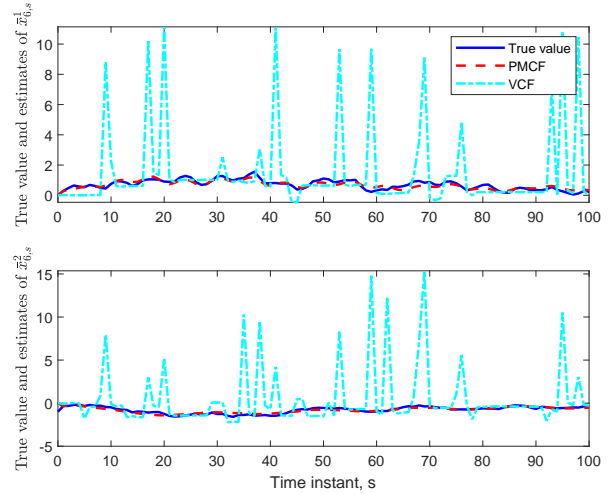


Fig. 11: True value and state estimates for node 6 in *Case 2*.

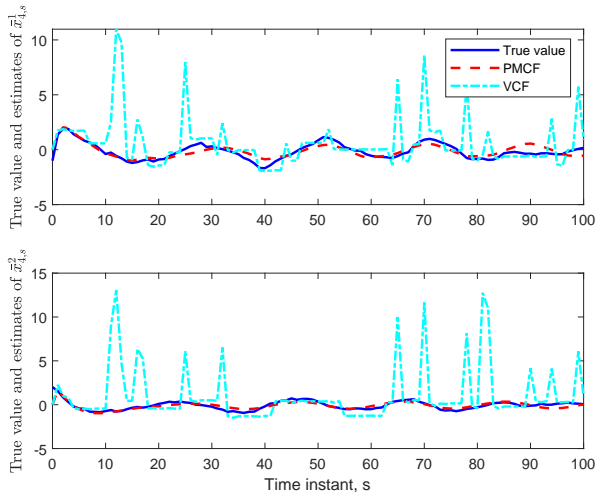


Fig. 9: True value and state estimates for node 4 in *Case 2*.

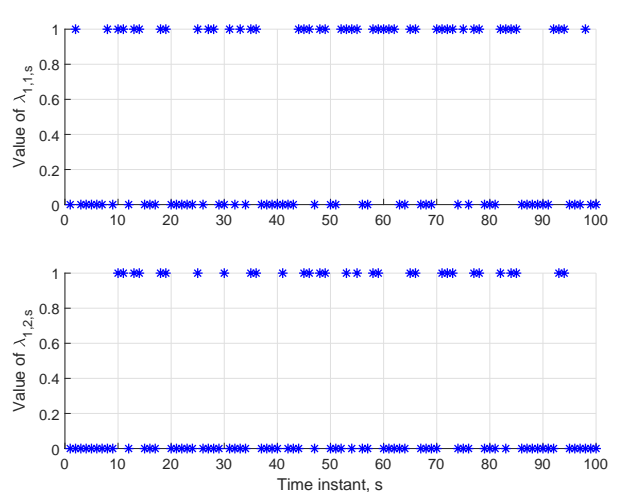


Fig. 12: Triggering instants for node 1 in *Case 2*.

correntropy criterion, which exhibits distinguishing advantages in terms of not only the robustness against non-Gaussian noises, but the resource saving as well. Finally, simulation results have been given to demonstrate the effectiveness of the proposed MCF algorithm. In the future, two possible research directions would be 1) investigating the influences of the node couplings onto the filtering performance; and 2) exploring the extensions of the developed results to other systems, such as the recurrent time-varying neural networks [31], the wireless sensor networks [53], and the multi-agent systems [51].

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